



A Simple Skew Angle Detection and Alleviation based on Readback Signal in Bit-Patterned Magnetic Recording

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MOTIVATION

✧ **Bit-patterned magnetic recording (BPMP)** can increase an areal density (AD) up to 4 Tb/in².

✧ **Skew angle (SA)** can change the relative placement of read/write elements on the slider, leading to a design issue in servo and write synchronization.

✧ In conventional systems, the SA can increase up to 35° (degrees) for inner and outer diameters [1].

✧ Without a SA detection and alleviation method, the system performance will dramatically degrade, particularly at high ADs.

✧ **Fig. 1(a)** displays a BPMP magnetic medium and an MR read head with SA effect, whereas **Fig. 1(b)** shows the impulse responses of upper, center, and lower tracks at 0° (degree) and 30° SA at AD of 3 Tb/in². Clearly, the impulse responses rely on the SA.

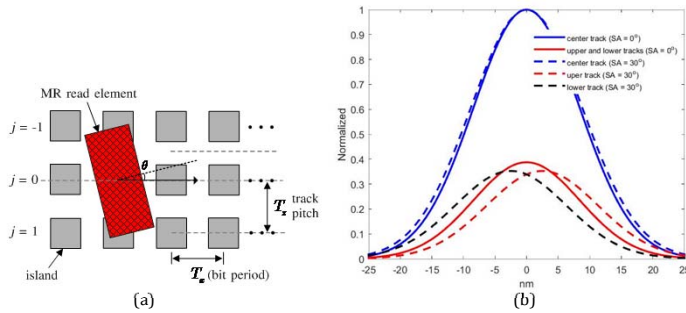


Fig. 1. (a) A BPMP medium and an MR head with SA effect and (b) the impulse responses of different tracks at AD = 3 Tb/in².

CHANNEL MODEL

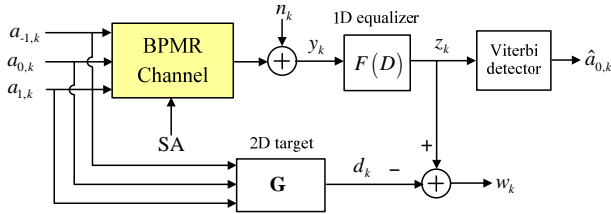


Fig. 2. A BPMP channel model with the SA effect.

✧ BPMP channel model with the SA effect

- Readback Signal

$$y_k = \sum_u \sum_v h_{u,v} a_{j-u,k-v} + n_k$$

where n_k is Additive white Gaussian noise.

- 2D Gaussian pulse response

$$H(z, x) = A \exp \left\{ -\frac{1}{2c^2} \left[\left(\frac{x_r}{PW_x} \right)^2 + \left(\frac{z_r}{PW_z} \right)^2 \right] \right\}$$

where $c = 1 / 2.3548$, PW_x is the PW₅₀ of the across-track pulse, and PW_z is the PW₅₀ of the along-track pulse.

where $x_r = x \cos(\theta) - z \sin(\theta)$, $z_r = x \sin(\theta) + z \cos(\theta)$, and θ is the skew angle in degree.

[1] Z. He, et al., MATEC Web of Conferences, 42 (2016).

[2] S. Koonkarnkhai, et al., ECTI Trans. on Comp. and Info. Tech., 6 (2012) 175 – 182.

Acknowledgments

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PROPOSED METHOD

✧ The SA detection

- **Fig. 3** shows the SA profiles of the channel coefficient $h_{-1,0}$ and the target coefficient $g_{-1,0}$ at 3, 3.5, and 4 Tb/in² for different SAs.

- We found that there is a **relationship** between $h_{-1,0}$ and $g_{-1,0}$.

- A simple SA detection method will utilize $g_{-1,0}$ to approximate the SA experienced in the system (i.e., $g_{-1,0} \approx h_{-1,0}$) by using the target and equalizer design based on an MMSE approach.

✧ The 2D BPMP channel matrix

$$H = \begin{bmatrix} h_{-1,-1} & h_{-1,0} & h_{-1,1} \\ h_{0,-1} & h_{0,0} & h_{0,1} \\ h_{1,-1} & h_{1,0} & h_{1,1} \end{bmatrix}$$

✧ The 2D target (G)

$$G = \begin{bmatrix} g_{-1,-1} & g_{-1,0} & g_{-1,1} \\ g_{0,-1} & g_{0,0} & g_{0,1} \\ g_{1,-1} & g_{1,0} & g_{1,1} \end{bmatrix}$$

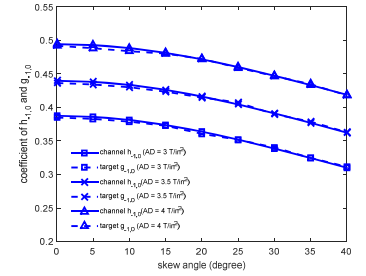


Fig. 3. The SA profile with respect to different channel and target coefficients.

✧ The SA alleviation

After the SA is detected, a pair of *flipped-cross-track symmetric* 2D target and 1D equalizer associated with the estimated SA is employed to alleviate the SA in data detection process.

✧ Target and Equalizer Design:

The target and its corresponding equalizer are designed by minimizing the mean-squared error between the equalizer output z_k and the target output d_k according to

$$E[e_k^2] = E[(z_k - d_k)^2] = E[(f^T y_k - g^T a_k)(f^T y_k - g^T a_k)^T]$$

✧ Design of a flipped-cross-track symmetric 2D target

Define: $g_{-1,-1} = g_{1,1}$, $g_{-1,0} = g_{1,0}$, $g_{-1,1} = g_{1,-1}$, $g = [g_{-1,-1} \ g_{-1,0} \ g_{-1,1} \ g_{0,-1} \ g_{0,0} \ g_{0,1} \ g_{1,-1} \ g_{1,0} \ g_{1,1}]^T$ is the column vector of the target, $f = [f_{-K} \ \dots \ f_0 \ \dots \ f_K]^T$ is the column vector of g_{-1} , the equalizer, $M = 2K + 1$ is the number of equalizer coefficients, L is the number of target coefficients, $R = E[y_k y_k^T]$ is an M -by- M auto-correlation matrix of y_k , $A = E[a_k a_k^T]$ is an L -by- L auto-correlation matrix of a_k , and $T = E[y_k a_k^T]$ is an M -by- L cross-correlation matrix of y_k and a_k , where $a_k = [a_{-1,k} \ a_{0,k} \ a_{1,k} \ a_{-1,k-1} \ a_{0,k-1} \ a_{1,k-2}]^T$ and $y_k = [y_{k+K} \ \dots \ y_k \ \dots \ y_{k-K}]^T$

MMSE: $E\{e_k^2\} = f^T R f + g^T A g - 2f^T T g$, where $I = [0 \ 0 \ 0 \ 0 \ 1 \ 0]^T$. Minimization process give:

$$\lambda = \frac{1}{I^T (A - T^T R^{-1} T) I} \quad g = \lambda (A - T^T R^{-1} T)^{-1} I \quad f = R^{-1} T g$$

SIMULATION RESULTS

✧ Parameter Setting

SNR = 10 log₁₀(1/σ²) in dB, **Equalizer** : 7-tap 1D equalizer, and **AD**: at 3.0 Tb/in², $T_x = T_z = 14.5$ nm. and 3.5 Tb/in², $T_x = T_z = 13.5$ nm.

✧ Define

- "Conv." is the system without SA detection and correction.
- "Prop." is the proposed system with flipped-cross-track symmetric 2D target.
- "Asym. 2D" is the proposed system with asymmetric 2D target [2].

✧ Conclusion

- The proposed method outperforms the conventional system as shown in **Fig. 4**.

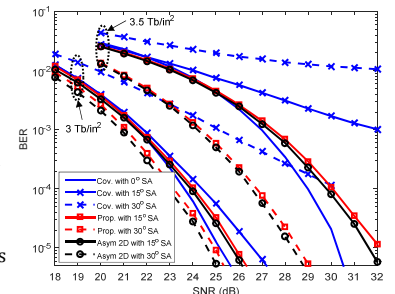


Fig. 4. Performance comparison at 3 and 3.5 Tb/in².

- The Viterbi detector used for the flipped-cross-track symmetric 2D target has lower complexity than an asymmetric 2D target.
- Although high SA provides better performance than low SA due to less ITI effect (not shown here), a large SA definitely causes mechanical problems during read/write processes.